

# 6.891

## Computer Vision and Applications

Prof. Trevor. Darrell

### Lecture 2:

- Linear Filters and Convolution (review)
- Fourier Transform (review)
- Sampling and Aliasing (review)

Readings: F&P Chapter 7.1-7.6

## Recap: Cameras, lenses, and calibration

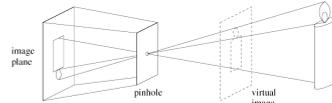
Last time:

- Camera models
- Projection equations
- Calibration methods

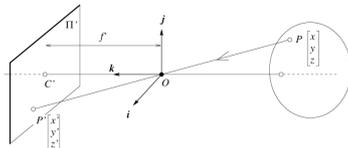
Images are projections of the 3-D world onto a 2-D plane...

## Recap: pinhole/perspective

Pinole camera model - box with a small hole in it:

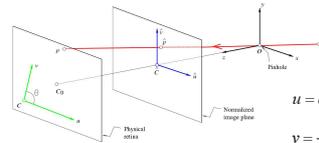


Perspective projection:



$$\begin{cases} x' = f \frac{x}{z} \\ y' = f \frac{y}{z} \end{cases}$$

## Recap: Intrinsic parameters



$$\begin{aligned} u &= \alpha \frac{x}{z} - \alpha \cot(\theta) \frac{y}{z} + u_0 \\ v &= \frac{\beta}{\sin(\theta)} \frac{y}{z} + v_0 \end{aligned}$$

Using homogenous coordinates, we can write this as:

$$\begin{pmatrix} u \\ v \\ 1 \end{pmatrix} = \frac{1}{z} \begin{pmatrix} \alpha & -\alpha \cot(\theta) & u_0 & 0 \\ 0 & \beta & v_0 & 0 \\ 0 & 0 & 1 & 0 \end{pmatrix} \begin{pmatrix} x \\ y \\ z \\ 1 \end{pmatrix}$$

or:

$$\vec{p} = \frac{1}{z} (K \ \vec{0}) \vec{P}_4$$

## Recap: Combining extrinsic and intrinsic calibration parameters

$$\vec{p} = \frac{1}{z} (K \ \vec{0}) \vec{P} \quad \text{Intrinsic}$$

$${}^c P = {}^c R {}^w P + {}^c O_w \quad \text{Extrinsic}$$

$$\vec{p} = \frac{1}{z} K ({}^c R \ {}^c O_w) \vec{P}$$

$$\vec{p} = \frac{1}{z} M \vec{P}$$

## Other ways to write the same equation

pixel coordinates

$$\vec{p} = \frac{1}{z} M \vec{P}$$

world coordinates

$$\begin{pmatrix} u \\ v \\ 1 \end{pmatrix} = \frac{1}{z} \begin{pmatrix} \cdot & m_1^T & \cdot & W_x \\ \cdot & m_2^T & \cdot & W_y \\ \cdot & m_3^T & \cdot & W_z \end{pmatrix} \begin{pmatrix} W_x \\ W_y \\ W_z \\ 1 \end{pmatrix}$$

$$\begin{cases} u = \frac{m_1 \cdot \vec{P}}{m_3 \cdot \vec{P}} \\ v = \frac{m_2 \cdot \vec{P}}{m_3 \cdot \vec{P}} \end{cases}$$

z is in the camera coordinate system, but we can solve for that, since  $1 = \frac{m_3 \cdot \vec{P}}{z}$ , leading to:

$$u = \frac{m_1 \cdot \bar{P}}{m_3 \cdot \bar{P}}$$

$$v = \frac{m_2 \cdot \bar{P}}{m_3 \cdot \bar{P}}$$

## Recap: Camera calibration



Stack all these measurements of  $i=1 \dots n$  points The Opti-CAL Calibration Target Image

$$(m_1 - u m_3) \cdot \bar{P}_i = 0$$

$$(m_2 - v m_3) \cdot \bar{P}_i = 0$$

into a big matrix:

$$\begin{pmatrix} P_1^T & 0^T & -u_1 P_1^T \\ 0^T & P_1^T & -v_1 P_1^T \\ \dots & \dots & \dots \\ P_n^T & 0^T & -u_n P_n^T \\ 0^T & P_n^T & -v_n P_n^T \end{pmatrix} \begin{pmatrix} m_1 \\ m_2 \\ m_3 \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 0 \end{pmatrix}$$

## Today

Review of early visual processing

- Linear Filters and Convolution
- Fourier Transform
- Sampling and Aliasing

*You should have been exposed to this material in previous courses; this lecture is just a (quick) review.*

Administrivia:

- sign-up sheet
- introductions

## What is image filtering?

- Modify the pixels in an image based on some function of a local neighborhood of the pixels.

10	5	3
4	5	1
1	1	7

Some function

→

	7	

Local image data
Modified image data <sup>9</sup>

## Linear functions

- Simplest: linear filtering.
  - Replace each pixel by a linear combination of its neighbors.
- The prescription for the linear combination is called the “convolution kernel”.

10	5	3
4	5	1
1	1	7

0	0	0
0	0.5	0
0	1	0.5

	7	

Local image data
kernel
Modified image data <sup>10</sup>

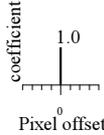
## Convolution

$$f[m, n] = I \otimes g = \sum_{k, l} I[m-k, n-l] g[k, l]$$

## Linear filtering (warm-up slide)



coefficient



Pixel offset

?

original

### Linear filtering (warm-up slide)

original      coefficient      Filtered (no change)

Pixel offset

13

### Linear filtering

original      coefficient      ?

Pixel offset

14

### shift

original      coefficient      shifted

Pixel offset

15

### Linear filtering

original      coefficient      ?

Pixel offset

16

### Blurring

original      coefficient      Blurred (filter applied in both dimensions)

Pixel offset

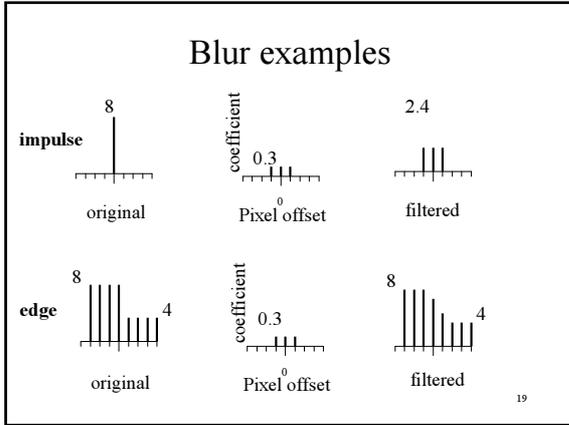
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### Blur examples

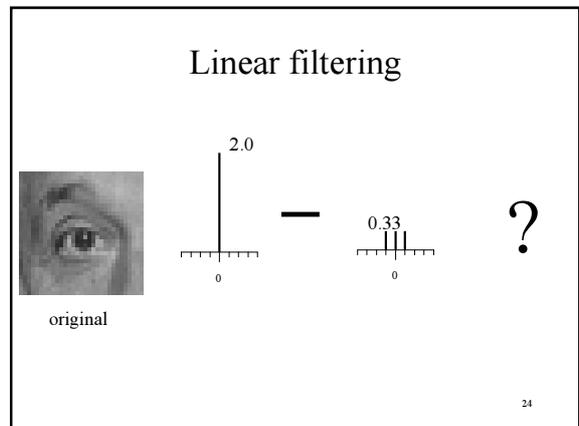
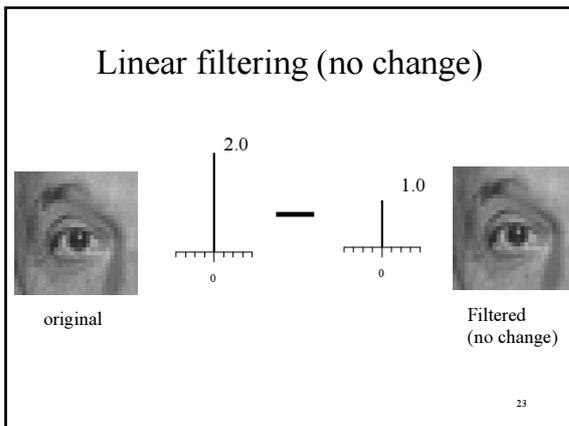
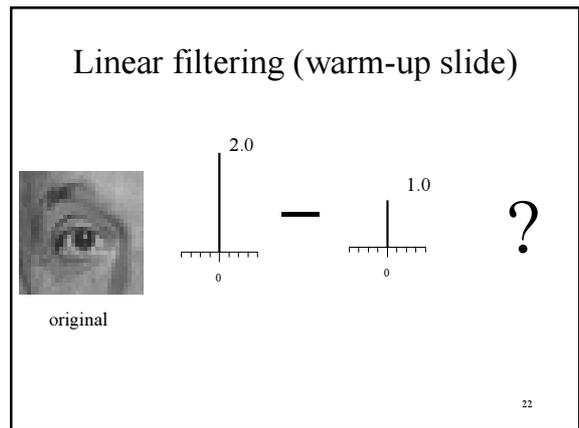
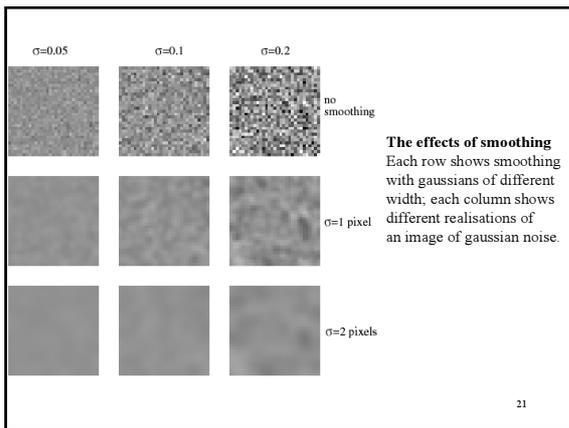
impulse      coefficient      2.4

original      Pixel offset      filtered

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- ### Smoothing reduces noise
- Generally expect pixels to “be like” their neighbours
    - surfaces turn slowly
    - relatively few reflectance changes
  - Generally expect noise processes to be independent from pixel to pixel
  - Implies that smoothing suppresses noise, for appropriate noise models
  - Scale
    - the parameter in the symmetric Gaussian
    - as this parameter goes up, more pixels are involved in the average
    - and the image gets more blurred
    - and noise is more effectively suppressed
- 20



(remember blurring)

original

coefficient

0.3

Pixel offset

Blurred (filter applied in both dimensions).

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Linear filtering

original

2.0

0

0.33

0

?

26

Sharpening

original

2.0

0

0.33

0

Sharpened original

27

Sharpening example

original

8

coefficient

1.7

-0.3

11.2

8

-0.25

Sharpened  
(differences are accentuated; constant areas are left untouched).

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Sharpening

before

after

29

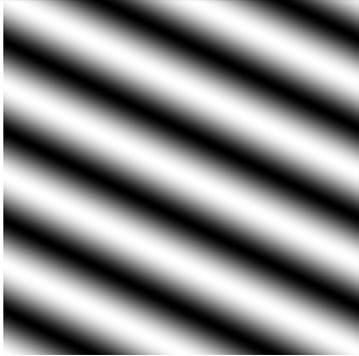
Gradients and edges

- Points of sharp change in an image are interesting:
  - change in reflectance
  - change in object
  - change in illumination
  - noise
- Sometimes called **edge points**
- General strategy
  - linear filters to estimate image gradient
  - mark points where gradient magnitude is particularly large wrt neighbours (ideally, curves of such points).

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To get some sense of what basis elements look like, we plot a basis element --- or rather, its real part --- as a function of  $x, y$  for some fixed  $u, v$ . We get a function that is constant when  $(ux+vy)$  is constant. The magnitude of the vector  $(u, v)$  gives a frequency, and its direction gives an orientation. The function is a sinusoid with this frequency along the direction, and constant perpendicular to the direction.



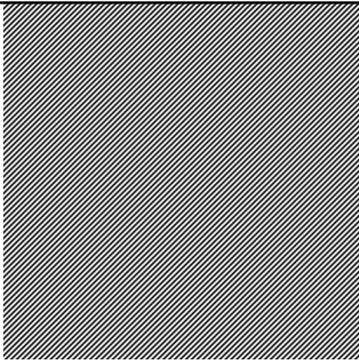
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Here  $u$  and  $v$  are larger than in the previous slide.



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And larger still...



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## Phase and Magnitude

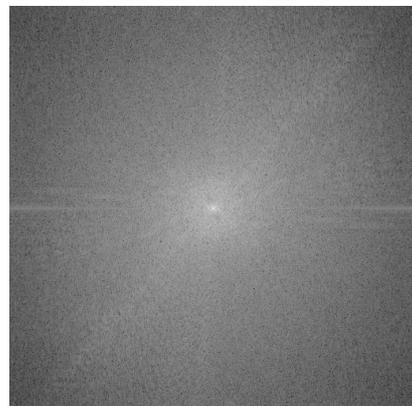
- Fourier transform of a real function is complex
  - difficult to plot, visualize
  - instead, we can think of the phase and magnitude of the transform
- Phase is the phase of the complex transform
- Magnitude is the magnitude of the complex transform
- Curious fact
  - all natural images have about the same magnitude transform
  - hence, phase seems to matter, but magnitude largely doesn't
- Demonstration
  - Take two pictures, swap the phase transforms, compute the inverse - what does the result look like?

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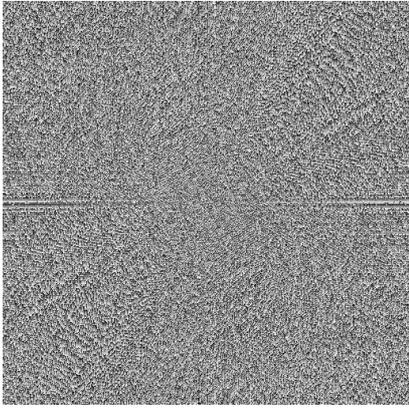
41

This is the magnitude transform of the cheetah pic



42

This is the phase transform of the cheetah pic

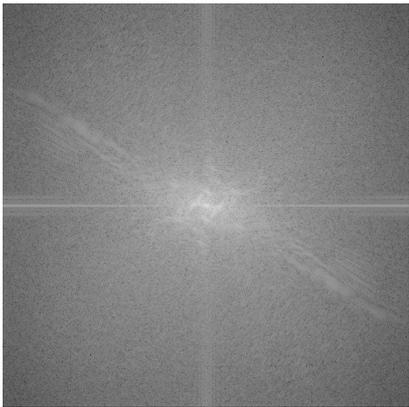


43



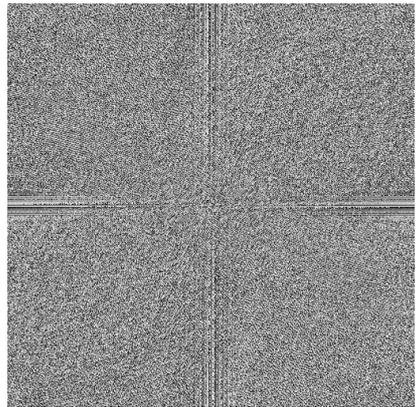
44

This is the magnitude transform of the zebra pic



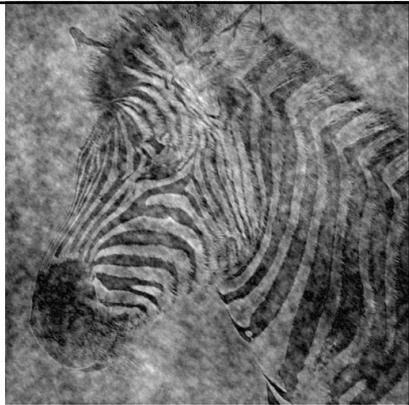
45

This is the phase transform of the zebra pic

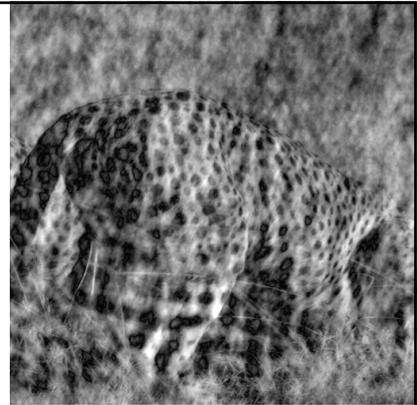


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Reconstruction with zebra phase, cheetah magnitude



Reconstruction with cheetah phase, zebra magnitude

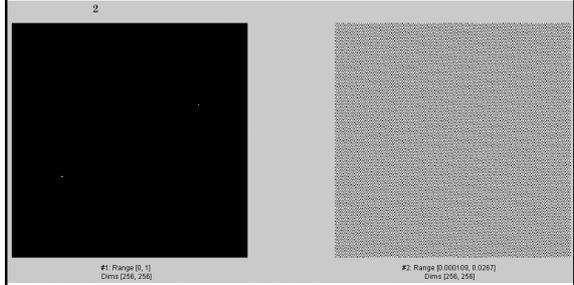


# Example image synthesis with fourier basis.

- 16 images

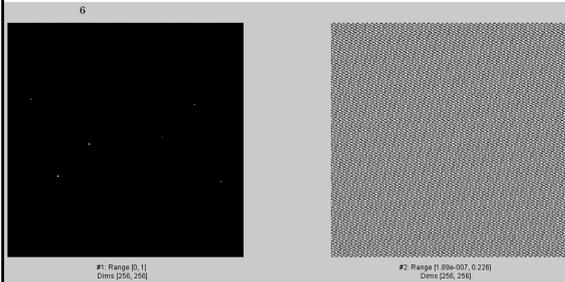
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2



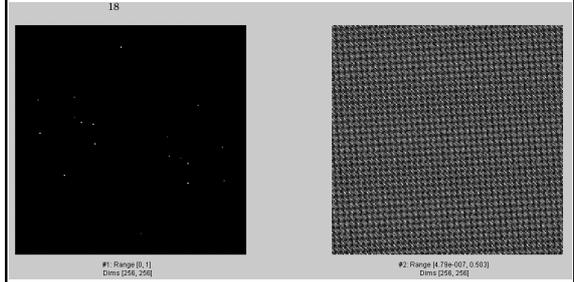
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6



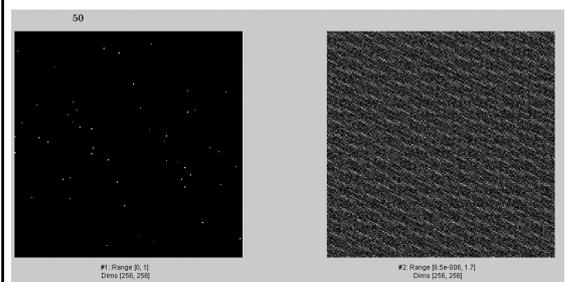
51

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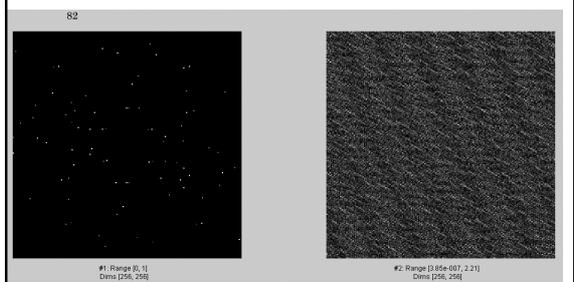
52

50

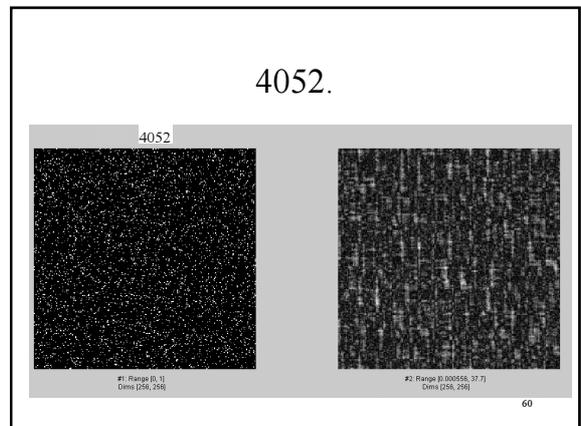
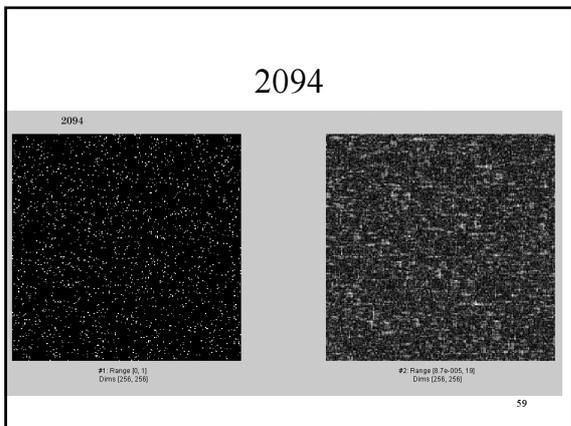
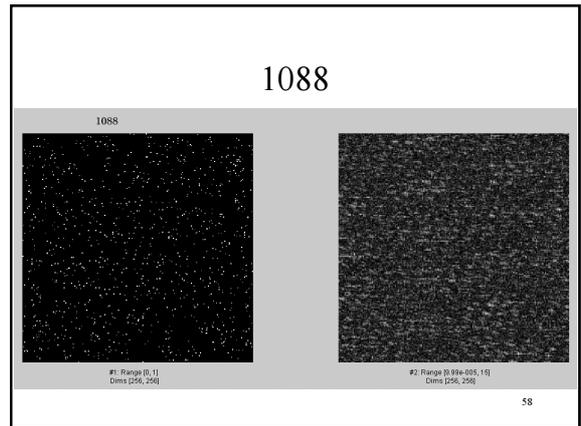
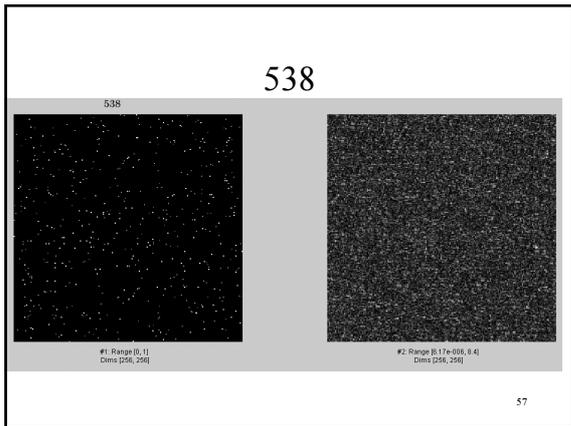
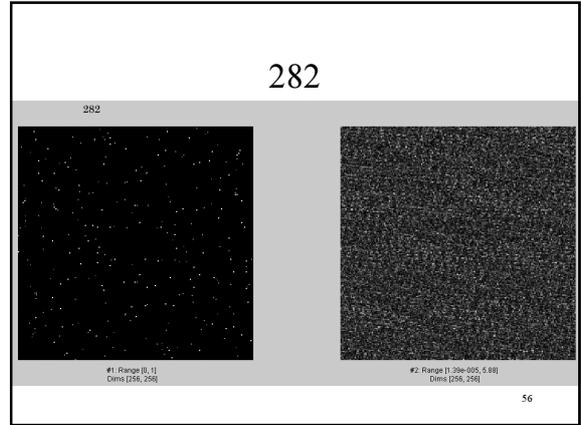
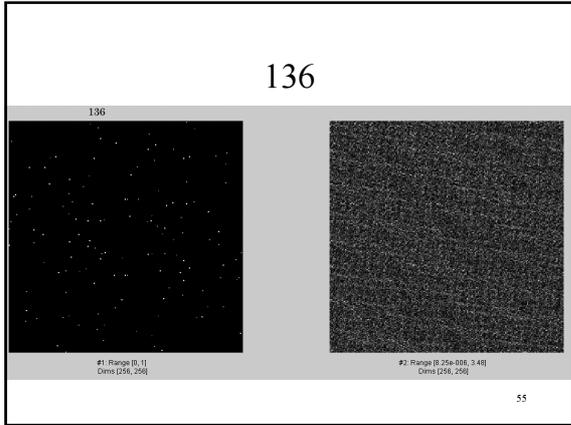


53

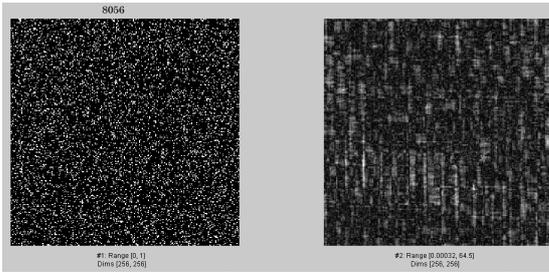
82



54

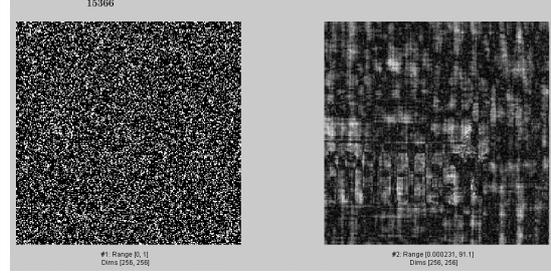


8056.



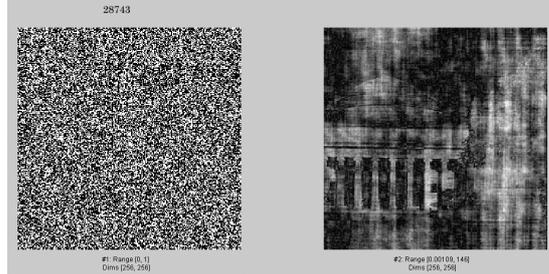
61

15366



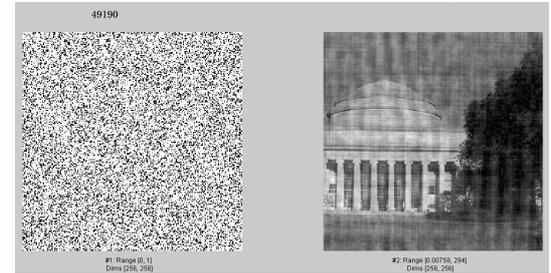
62

28743



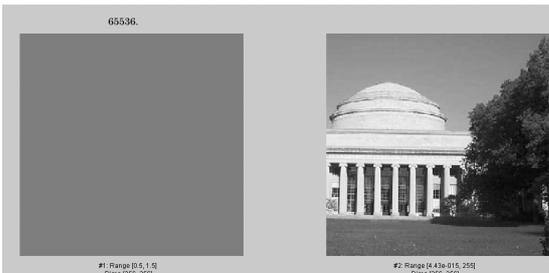
63

49190.



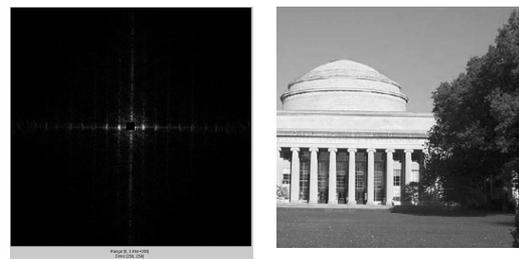
64

65536.



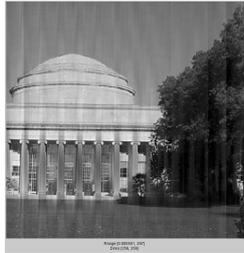
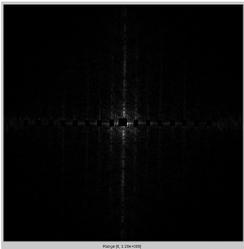
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Fourier transform magnitude



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## Masking out the fundamental and harmonics from periodic pillars



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Name as many functions as you can that retain that same functional form in the transform domain

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TABLE 7.1 A variety of functions of two dimensions and their Fourier transforms. This table is used in two directions (with appropriate substitutions for  $u, v$  and  $x, y$ ) because the Fourier transform of a function is the function. Observant readers may suspect that the results on infinite sums of  $\delta$  functions contradict the linearity of Fourier transforms. By careful inspection of limits, it is possible to show that they do not (see, e.g., Bracewell, 1978). Observant readers may also have noted that an expression for  $\mathcal{F}\{\frac{1}{r^2}\}$  can be obtained by combining two lines of this table.

Function	Fourier transform
$g(x, y)$	$\iint_{-\infty}^{\infty} g(x, y)e^{-i2\pi(ux+vy)} dx dy$
$\iint_{-\infty}^{\infty} \mathcal{F}\{g(x, y)\}u, v\}e^{i2\pi(ux+vy)} du dv$	$\mathcal{F}\{g(x, y)\}(u, v)$
$\delta(x, y)$	1
$\frac{\delta(x, y)}{xy}$	$u\mathcal{F}\{f(u, v)\}$
$0.5\delta(x+a, y) + 0.5\delta(x-a, y)$	$\cos 2\pi ax$
$e^{-i2\pi(x^2+y^2)}$	$e^{-\pi^2(u^2+v^2)}$
$\text{hsv}(x, y)$	$\frac{\sin \pi u \sin \pi v}{u v}$
$f(ax, by)$	$\frac{\mathcal{F}\{f\}(u/a, v/b)}{ab}$
$\sum_{l=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta(x-l, y-m)$	$\sum_{l=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta(u-l, v-m)$
$\mathcal{F}\{f+g\}(x, y)$	$\mathcal{F}\{f\}+\mathcal{F}\{g\}(u, v)$
$\mathcal{F}\{f-a, y-b\}$	$e^{-i2\pi(ax+by)}\mathcal{F}\{f\}$
$f(x \cos \theta - y \sin \theta, x \sin \theta + y \cos \theta)$	$\mathcal{F}\{f\}(u \cos \theta - v \sin \theta, u \sin \theta + v \cos \theta)$

Forsyth & Ponce

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## Discrete-time, continuous frequency Fourier transform

Many sequences can be represented by a Fourier integral of the form

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega, \quad (2.133)$$

where

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}. \quad (2.134)$$

Oppenheim, Schafer and Buck, Discrete-time signal processing, Prentice Hall, 1999

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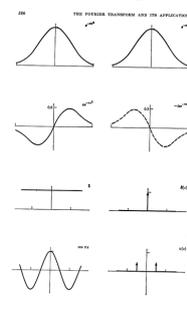
## Discrete-time, continuous frequency Fourier transform pairs

TABLE 2.3 FOURIER TRANSFORM PAIRS

Sequence	Fourier Transform
1. $\delta[n]$	1
2. $\delta[n-n_0]$	$e^{-j\omega n_0}$
3. 1 $(-\infty < n < \infty)$	$\sum_{k=-\infty}^{\infty} 2\pi \delta(\omega + 2\pi k)$
4. $a^n u[n]$ $( a  < 1)$	$\frac{1}{1 - ae^{-j\omega}}$
5. $u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi \delta(\omega + 2\pi k)$
6. $(n+1)a^n u[n]$ $( a  < 1)$	$\frac{1}{(1 - ae^{-j\omega})^2}$
7. $\frac{r^n \sin \omega_0 n + 1}{\sin \omega_0} u[n]$ $( r  < 1)$	$\frac{1}{1 - 2r \cos \omega_0 e^{-j\omega} + r^2 e^{-j2\omega}}$
8. $\frac{\sin \omega_0 n}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1, &  \omega  \leq \omega_0 \\ 0, & \omega_0 <  \omega  \leq \pi \end{cases}$
9. $x[n] = \begin{cases} 1, & 0 \leq n \leq M \\ 0, & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
10. $e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi \delta(\omega - \omega_0 + 2\pi k)$
11. $\cos(\omega_0 n + \phi)$	$\sum_{k=-\infty}^{\infty} [\pi e^{j\phi} \delta(\omega - \omega_0 + 2\pi k) + \pi e^{-j\phi} \delta(\omega + \omega_0 + 2\pi k)]$

Oppenheim, Schafer and Buck, Discrete-time signal processing, Prentice Hall, 1999

## Bracewell's pictorial dictionary of Fourier transform pairs



Bracewell, The Fourier Transform and its Applications, McGraw Hill 1978

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Bracewell's pictorial dictionary of Fourier transform pairs

Bracewell, The Fourier Transform and its Applications, McGraw Hill 1978

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Bracewell's pictorial dictionary of Fourier transform pairs

Bracewell, The Fourier Transform and its Applications, McGraw Hill 1978

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Bracewell's pictorial dictionary of Fourier transform pairs

Bracewell, The Fourier Transform and its Applications, McGraw Hill 1978

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Why is the Fourier domain particularly useful?

- It tells us the effect of linear convolutions.

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Fourier transform of convolution

Consider a (circular) convolution of  $g$  and  $h$

$$f = g \otimes h$$

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Fourier transform of convolution

$$f = g \otimes h$$

Take DFT of both sides

$$F[m, n] = DFT(g \otimes h)$$

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### Fourier transform of convolution

$$f = g \otimes h$$

$$F[m, n] = DFT(g \otimes h)$$

Write the DFT and convolution explicitly

$$F[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] h[k, l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)}$$

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### Fourier transform of convolution

$$f = g \otimes h$$

$$F[m, n] = DFT(g \otimes h)$$

$$F[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] h[k, l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)}$$

Move the exponent in

$$= \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)} h[k, l]$$

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### Fourier transform of convolution

$$f = g \otimes h$$

$$F[m, n] = DFT(g \otimes h)$$

$$F[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] h[k, l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)}$$

$$= \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)} h[k, l]$$

Change variables in the sum

$$= \sum_{\mu=-k}^{M-k-1} \sum_{\nu=-l}^{N-l-1} \sum_{k,l} g[\mu, \nu] e^{-\pi j \left( \frac{(k+\mu)m}{M} + \frac{(l+\nu)n}{N} \right)} h[k, l]$$

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### Fourier transform of convolution

$$f = g \otimes h$$

$$F[m, n] = DFT(g \otimes h)$$

$$F[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] h[k, l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)}$$

$$= \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)} h[k, l]$$

$$= \sum_{\mu=-k}^{M-k-1} \sum_{\nu=-l}^{N-l-1} \sum_{k,l} g[\mu, \nu] e^{-\pi j \left( \frac{(k+\mu)m}{M} + \frac{(l+\nu)n}{N} \right)} h[k, l]$$

Perform the DFT (circular boundary conditions)

$$= \sum_{k,l} G[m, n] e^{-\pi j \left( \frac{km}{M} + \frac{ln}{N} \right)} h[k, l]$$

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### Fourier transform of convolution

$$f = g \otimes h$$

$$F[m, n] = DFT(g \otimes h)$$

$$F[m, n] = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] h[k, l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)}$$

$$= \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} \sum_{k,l} g[u-k, v-l] e^{-\pi j \left( \frac{um}{M} + \frac{vn}{N} \right)} h[k, l]$$

$$= \sum_{\mu=-k}^{M-k-1} \sum_{\nu=-l}^{N-l-1} \sum_{k,l} g[\mu, \nu] e^{-\pi j \left( \frac{(k+\mu)m}{M} + \frac{(l+\nu)n}{N} \right)} h[k, l]$$

$$= \sum_{k,l} G[m, n] e^{-\pi j \left( \frac{km}{M} + \frac{ln}{N} \right)} h[k, l]$$

Perform the other DFT (circular boundary conditions)

$$= G[m, n] H[m, n]$$

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### Analysis of our simple filters

85

### Analysis of our simple filters



coefficient



Pixel offset



original      Pixel offset      Filtered  
(no change)

$$F[m, n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} f[k, l] e^{-j\pi \left( \frac{km}{M} + \frac{ln}{N} \right)}$$

$$= 1 \quad \frac{1.0}{\text{constant}}$$



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### Analysis of our simple filters



coefficient



Pixel offset



original      Pixel offset      shifted

$$F[m, n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} f[k - \delta, l] e^{-j\pi \left( \frac{km}{M} + \frac{ln}{N} \right)}$$

$$= e^{-j\pi \frac{\delta m}{M}}$$



Constant magnitude, linearly shifted phase

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### Analysis of our simple filters



coefficient



Pixel offset



original      Pixel offset      blurred

$$F[m, n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} f[k, l] e^{-j\pi \left( \frac{km}{M} + \frac{ln}{N} \right)}$$

$$= \frac{1}{3} \left( 1 + 2 \cos \left( \frac{\pi m}{M} \right) \right)$$



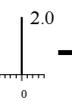
Low-pass filter

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### Analysis of our simple filters



coefficient



Pixel offset



original      Pixel offset      sharpened

$$F[m, n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} f[k, l] e^{-j\pi \left( \frac{km}{M} + \frac{ln}{N} \right)}$$

$$= 2 - \frac{1}{3} \left( 1 + 2 \cos \left( \frac{\pi m}{M} \right) \right)$$

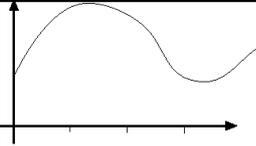


high-pass filter

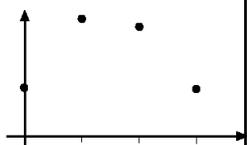
89

## Sampling and aliasing

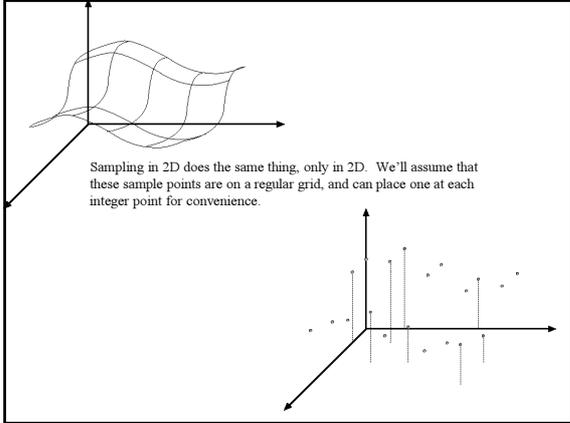
90



Sampling in 1D takes a continuous function and replaces it with a vector of values, consisting of the function's values at a set of sample points. We'll assume that these sample points are on a regular grid, and can place one at each integer for convenience.



90



## A continuous model for a sampled function

- We want to be able to approximate integrals sensibly
- Leads to
  - the delta function
  - model on right

$$\text{Sample}_{2D}(f(x, y)) = \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} f(x, y) \delta(x-i, y-j)$$

$$= f(x, y) \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} \delta(x-i, y-j)$$

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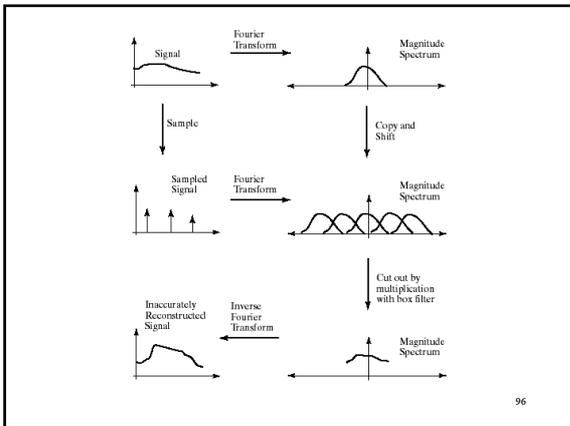
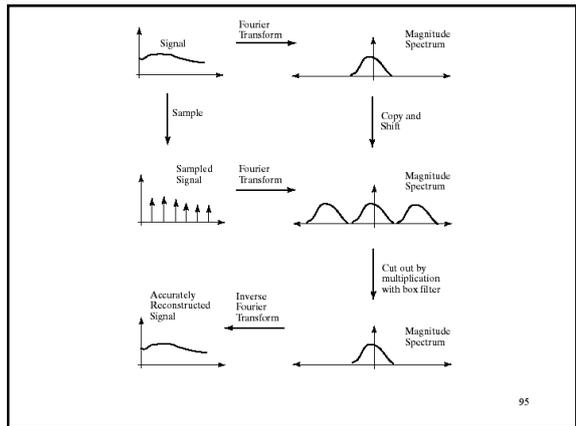
## The Fourier transform of a sampled signal

$$F(\text{Sample}_{2D}(f(x, y))) = F\left(f(x, y) \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} \delta(x-i, y-j)\right)$$

$$= F(f(x, y)) * F\left(\sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} \delta(x-i, y-j)\right)$$

$$= \sum_{i=-\infty}^{\infty} \sum_{j=-\infty}^{\infty} F(u-i, v-j)$$

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## Aliasing

- Can't shrink an image by taking every second pixel
- If we do, characteristic errors appear
  - In the next few slides
  - Typically, small phenomena look bigger; fast phenomena can look slower
  - Common phenomenon
    - Wagon wheels rolling the wrong way in movies
    - Checkerboards misrepresented in ray tracing
    - Striped shirts look funny on colour television

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Resample the checkerboard by taking one sample at each circle. In the case of the top left board, new representation is reasonable. Top right also yields a reasonable representation. Bottom left is all black (dubious) and bottom right has checks that are too big.

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Constructing a pyramid by taking every second pixel leads to layers that badly misrepresent the top layer

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## Smoothing as low-pass filtering

- The message of the FT is that high frequencies lead to trouble with sampling.
- Solution: suppress high frequencies before sampling
  - multiply the FT of the signal with something that suppresses high frequencies
  - or convolve with a low-pass filter
- A filter whose FT is a box is bad, because the filter kernel has infinite support
- Common solution: use a Gaussian
  - multiplying FT by Gaussian is equivalent to convolving image with Gaussian.

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Sampling without smoothing. Top row shows the images, sampled at every second pixel to get the next; bottom row shows the magnitude spectrum of these images.

256x256   128x128   64x64   32x32   16x16

101

Sampling with smoothing. Top row shows the images. We get the next image by smoothing the image with a Gaussian with sigma 1 pixel, then sampling at every second pixel to get the next; bottom row shows the magnitude spectrum of these images.

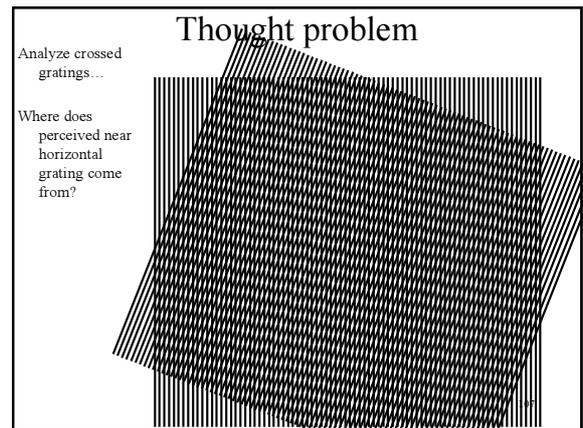
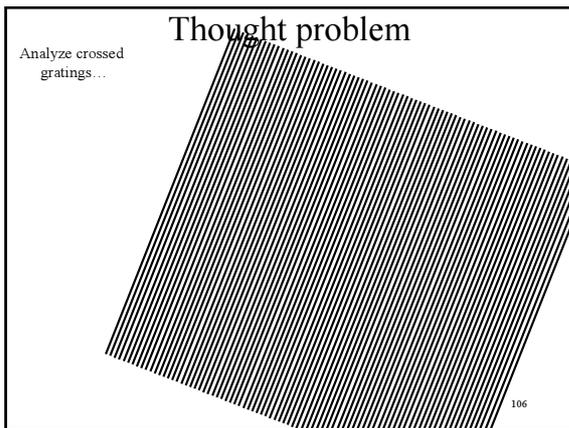
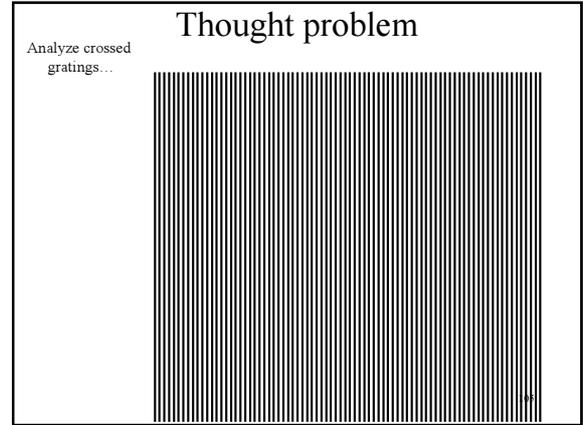
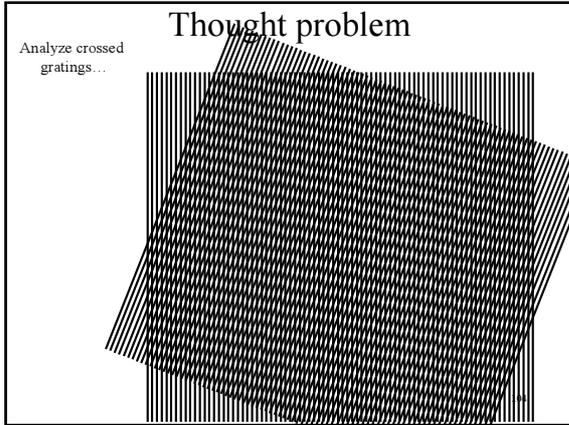
256x256   128x128   64x64   32x32   16x16

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Sampling with smoothing. Top row shows the images. We get the next image by smoothing the image with a Gaussian with sigma 1.4 pixels, then sampling at every second pixel to get the next; bottom row shows the magnitude spectrum of these images.

256x256   128x128   64x64   32x32   16x16

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What is a good representation for image analysis?

- Fourier transform domain tells you “what” (textural properties), but not “where”.
- Pixel domain representation tells you “where” (pixel location), but not “what”.
- Want an image representation that gives you a local description of image events—what is happening where.